

Simulation of 4-Leg Inverter based APF for PV based System with Effective Controller

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Abstract: Today's rising electricity consumption need more robust power plants. As a result, we're running out of fossil fuels, which increases pollution. As a result, we are shifting to renewable energy sources, which are better for the environment and save us money in the long run. Power electronic converters are being used more often in distribution networks to link RES. Solar radiation is one of the most widely distributed renewable energy sources worldwide. PV Cell is used to convert solar energy into usable electricity. Consequently, the PV Cell is applied to the converter and disseminated to the matrix later being thus constructed. Numerous controlling issues arise from the non-linear features of the load, despite the fact that numerous methods have been provided for modelling and engineering the PV cell and its point of interaction to the matrix. This exploration presents another control strategy for advancing the exhibition of these framework connecting inverters in 3-stage 4-wire appropriation organizations. The inverter is customized to fill a few needs by including a functioning power channel. To adapt to flow unevenness, load flow music, receptive power interest, and burden impartial flow, the inverter might act as a power converter while infusing power created from RES into the lattice, and as a shunt dynamic power channel while adjusting the framework's current. It is feasible to do every one of these activities independently or at the same time. An inverter that connection points with the lattice and a 3-stage, 4-wire straight/non-direct lopsided burden at the purpose in like manner coupling might be made to appear to the network as a fair direct burden by using this kind of control. Extensive MATLAB/Simulink simulation experiments prove the efficacy of this novel control strategy.

Index Terms: Controlled current, active power filtering, four-leg converters, and predictive regulation.

1. INTRODUCTION

In order to keep up with the rising demand for electricity, electric utilities and electricity consumers are becoming more anxious. It is estimated that the combustion of fossil fuels provides 75% of the world's energy needs. However, the need to seek for other energy sources has arisen in light of rising levels of air pollution, global warming concerns, the depletion of fossil fuels, and the rising expense of their use. Over the last decade, several nations have placed a major emphasis on developing renewable energy sources to meet their growing need for electricity. Market deregulation and government subsidies have both contributed to the expansion of the renewable energy industry.

Distributed generation (DG) refers to the integration of RES at the distribution level. Utility companies worry that instability, voltage control, and power-quality (PQ) difficulties may arise owing to the widespread use of intermittent RES in distribution networks. To guarantee the security, dependability, and efficacy of the whole

network, DG systems must conform to stringent technological and legal frameworks. Enhanced system operation and power quality (PQ) at the PCC are now possible because to the advent of power electronics and digital control technology, which allows for active management of the DG systems.

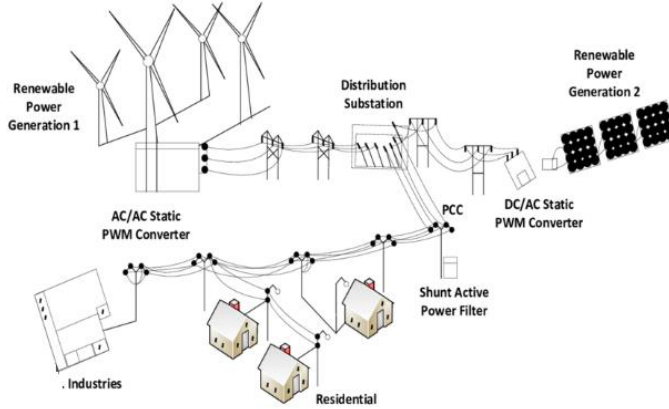


Fig. 1. Shunt active power filter used in a standalone hybrid power generating system.

Tragically, PCC's weighty dependence on power hardware based gear and non-direct loads bring about consonant flows that might debase power quality [1, 2]. In a distributed system, the intermittent RES are often interfaced using current regulated voltage source inverters. There have been recent proposals for control techniques for grid-connected inverters that use PQ solution. To remove the harmonic current, the inverter in [3] acts as an active inductor at a predetermined frequency. However, real-time, accurate calculations of network inductance may be challenging and may degrade control performance. Active conductance in the form of a shunt active filter is presented in [4] as a means of mitigating distribution network harmonics. In [5], the authors offer a -theory-based technique for controlling inverters that interface with renewable energy sources. In this method, monitoring both the heap current and the inverter current is vital for symphonious remuneration of the heap current.

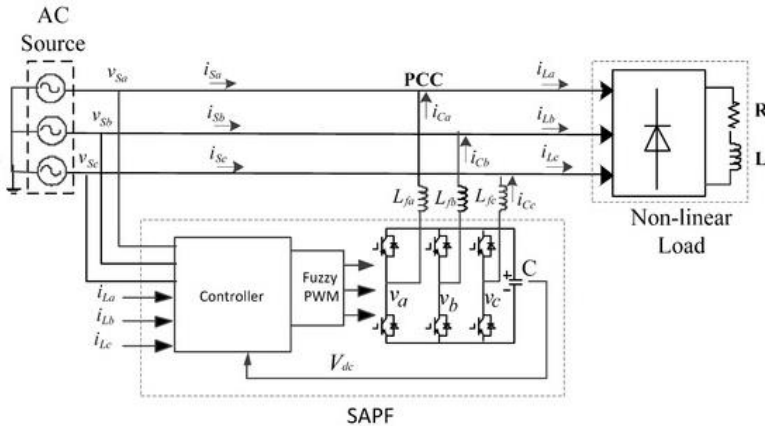


Fig. 2. The proposed shunt active power filter features an analogous circuit in three phases.

Harmonics in the current drawn from a non-linear load may cause voltage harmonics, which in turn may wreak havoc on the power quality of the grid as a whole. Distribution level harmonics and load imbalance may be mitigated with the use of active power filters (APF). This causes a rise in hardware expenses. In contrast, the standard inverter used to interface renewable energy with the grid has been enhanced with APF functionality in this article. The objective here is to make the most of the inverter's rating, which is often wasted owing to the unreliable output of RES.

This article details the design process for the proposed predictive control method, as well as the mathematics behind the 4Leg-VSI's mathematical model. The current reference generator that was chosen to be used in the unique power divert is additionally portrayed exhaustively. At last, modeling data and experimental findings

from a 2 kVA system are used to illustrate the efficacy of the effectiveness of the control plan that goes along with the active power filter that is being proposed adjustment.

II. FOUR-LEG CONVERTER MODEL

A example power distribution system that makes use of renewable energy sources is seen in Fig. 1. There are several distinct power generators and loads in this system. The majority of power used in homes and small businesses comes from renewable sources like wind and sunshine. Both ac/ac and dc/ac static PWM converters and battery banks are used for voltage change and long stretch energy capacity. In order to get the most power out of renewable sources like wind and sun, these converters use a technique called maximum power point tracking. Because of the unpredictable and erratic nature of electrical energy consumption, it very well may be either single-or three-stage, changed or inconsistent, straight or nonlinear. A functioning power channel is associated in lined up at the common connection to eliminate reactive power, unbalanced current, and harmonic distortion. It is composed of the parts shown in Fig. 2, which are an electrolytic capacitor, a four-leg beat width change converter, and a first-demand yield grow channel. The heap's Z_L , the result swell channel's Z_f , and the power framework's Z_s are the equivalent impedances. All taken into account in this circuit.

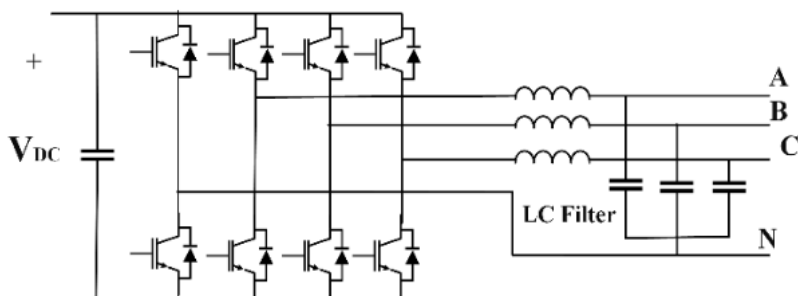


Fig. 3. Two-tier, four-legged PWM-VSI structure.

In Fig. 3 we see the architecture of a four-legged PWM converter. With the fourth leg associated with the unbiased transport, this converter architecture functions similarly to a standard three-phase converter. Current imbalance compensation is possible with the expansion of the fourth leg, which raises the all out number of exchanging states.

It is possible to define the voltage in any leg ‘x’ of the converter concerning exchanging states, with referenceto the voltage at the neutral point (n):

$$V_{xn} = (S_x - S_n) V_{dc}, x = u, v, w, n \quad \dots\dots(1)$$

From Fig.2's analogous circuit, we can deduce the filter's mathematical model, which is as follows:

$$V_0 = V_{xn} - R_{eq} i_0 - L_{eq} di_0/dt \quad \dots\dots(2)$$

Where, R_{eq} and L_{eq} are the required and needed Thevenin’s impedances at the converter's outcome terminals, independently, and Z_{eq} is the result voltage.

III. DIGITAL PREDICTIVE CURRENT CONTROL

Fig. 4 is a block diagram that explains the proposed digital predictive current control technique. Since this control method is basically an advancement calculation, it should be programmed into a microprocessor. Because of this, discrete mathematics is required to create the analysis so that factors like time delays and

Approximations may be taken into account.

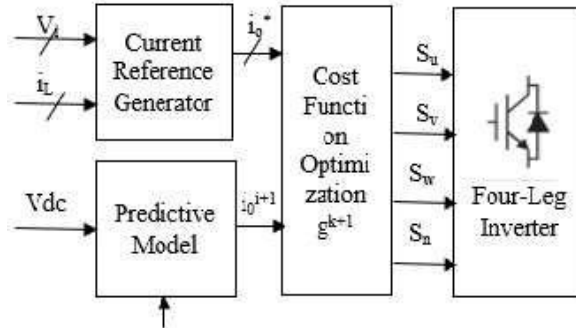


Fig. 4. Conceptual Block Diagram for the Predictive Digital Current Control

Prescient control is recognized fundamentally by its dependence on the framework model for making expectations about the future way of behaving of the controlled factors. With this data in hand, the controller can make an informed decision on the ideal changing state to apply to the power converter. As can be seen in Fig. 4, the predictive control method only requires three primary components, making it simple to construct and comprehend.

IV. CURRENT REFERENCE GENERATION

The ongoing reference signals for the dynamic power channel are gotten utilizing a dq-based current reference generator approach. This technique gives a fast and exact method for following signs. This feature prevents voltage fluctuations that degrade the performance of correction due to the current reference signal [6]. As illustrated in Fig. 5, the ongoing reference signals are gotten from the matching burden flows. To account for receptive power, current sounds, and current unevenness, this part decides the reference signal streams expected by the converter. The removal power factor ($\sin(L)$) and the heap's greatest absolute consonant twisting (THD (L)) decide the evident power expected by the dynamic power channel as illustrated.

$$\frac{SAPF}{SL} = \frac{\sqrt{\sin\phi(L)+THD(L)}x^2}{\sqrt{1+THD(L)}x^2} \dots\dots(3)$$

Where the greatest harmonic current that may be compensated for is included in the THD (L) value, where L is characterized as two times the inspecting recurrence f_s . A big part of the converter trading repeat is the most noteworthy recurrence of the symphonious current component that can be corrected. Since the dq-based technique works in a pivoting reference outline, the $\sin(\omega t)$ and $\cos(\omega t)$ signals must be applied to the measured currents to account for the rotation. dq-transformation involves 90-degree phase shifting of the q current component and synchronization of the d current part with the stage to-unprejudiced system voltage. Reference signals $\sin(\omega t)$ and $\cos(\omega t)$ are synchronized using a planned reference outline (SRF) stage locked circle (PLL). In any event, when the framework voltage is very contorted, the SRF - PLL actually delivers a clean sinusoidal waveform.

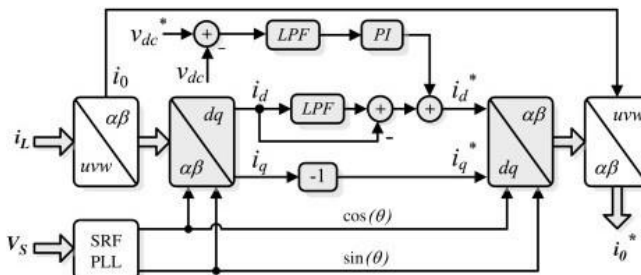


Fig.5. Block schematic of a dq-based current reference generator.

By avoiding stage voltage unbalancing, music (i.e., under 5% and 3% in fifth and seventh, independently), and offset conveyed by nonlinear weight conditions and assessment bungles [7], SRF - PLLs eliminate following mistakes. Real currents $i_{Lx}(t)$ ($x=u, v, \text{ and } w$) and the dq components of those currents (i_d and i_q) are related to one another, as shown in Equation (4) & (5):

$$\frac{i_d}{i_q} = \sqrt{\frac{2}{3}} \begin{bmatrix} \sin wt & \cos wt \\ -\cos wt & \sin wt \end{bmatrix} \begin{bmatrix} 1 & -\frac{1}{2} \\ 0 & \frac{\sqrt{3}}{2} \end{bmatrix} \begin{bmatrix} -\frac{1}{2} & -\frac{1}{2} \\ \frac{\sqrt{3}}{2} & -\frac{\sqrt{3}}{2} \end{bmatrix} \begin{bmatrix} i_{Lu} \\ i_{Lv} \\ i_{Lw} \end{bmatrix} \dots\dots(4)$$

As illustrated in the following diagram, The load's neutral current is balanced by infusing a similar immediate worth got from the stage flows but with a 180-degree phase shift:

$$i_{on}^* = -(i_u + i_v + i_w) \dots\dots(5)$$

The utilization of a direct regulator in the dc-voltage control circle is a colossal benefit of the dq-based current reference generator framework. However, under imbalanced operating circumstances, a second solicitation symphonious part is molded in i_d and knowledge level due to the dq-based current reference frame methodology used to fabricate the continuous reference. This harmonic's amplitude is proportional to the percentage of imbalanced load current (which is represented by the ratio of the negative progression current $i_{L,2}$ to the positive gathering current $i_{L,1}$). Since the second-request consonant is still present in i_d and i_q , a third harmonic is produced when the reference current is switched back over completely to the abc outline. In Fig. 6 we see the association between the percent of weight current lopsidedness and the percent of third consonant system current. Since the pile current misses the mark on third consonant, the unique power channel contributes one to the power grid.

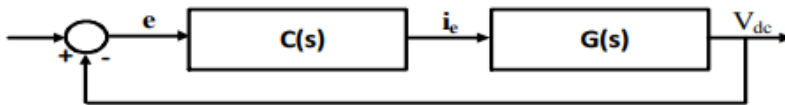


Fig. 6. Block schematic of a DC-voltage controller.

A standard PI regulator is utilized to manage the dc-voltage converter. Since the expense capability (6) is created involving only modern references to forego the utilization of weighting variables, this is a crucial problem in the evaluation. These parameters are usually gained by experimentation, and their definition becomes murky when a variety of operational circumstances must be met. Moreover, the current transient response is unaffected by the sluggish the electrolytic capacitor's dynamic response to the voltage that crosses it. To this end the PI regulator is a feasible choice to the dc-voltage control since it is both simple to execute and extremely powerful. The dc voltage will stay steady (with a base worth of 6 volts (rms)) until the dynamic power consumed by the converter dips under an edge beneath which it can't make up for its misfortunes.

V. PHOTOVOLTAIC SYSTEM

Solar energy may be transformed into usable electricity via the use of a photovoltaic (PV) system. The PV cell is the primary component of a PV system. Arrays may be created from groups of cells. Small loads, including lights and DC motors, may be powered directly from a PV device's terminal voltage and current, or the device can be connected to the grid through the appropriate energy conversion equipment.

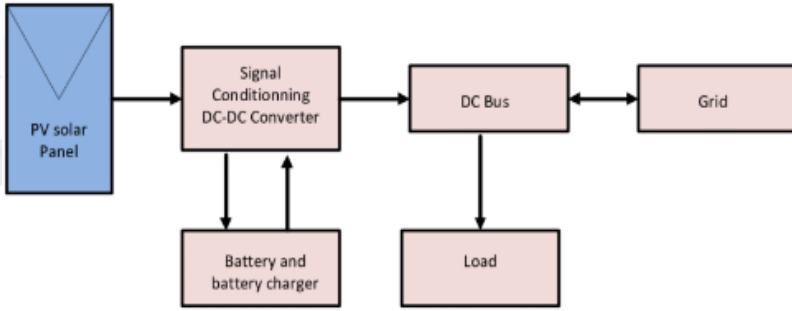


Fig. 7. A block schematic of a photovoltaic system

There are three primary components to this photovoltaic system: the PV module, the system's equilibrium, and the load. The charger, battery, and inverter make up the bulk of this system's energy balance. Block graph of the PV framework is displayed in Fig. 7.

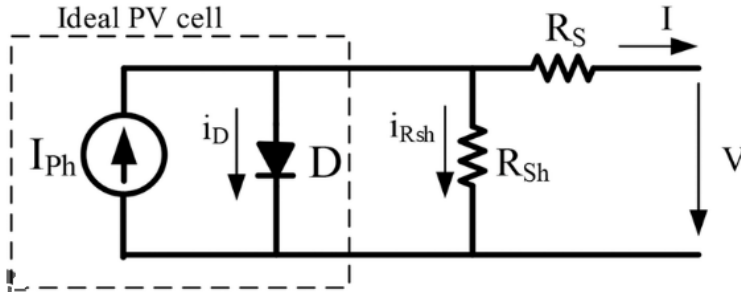


Fig.8. Practical PV device

To put it simply, a photovoltaic cell is a semiconductor diode with its p-n intersection presented to light. Different semiconductor materials and techniques are used in the production of photovoltaic cells. When a cell is exposed to light, it produces charge carriers, which, when combined with a short circuit, create an electric current I .

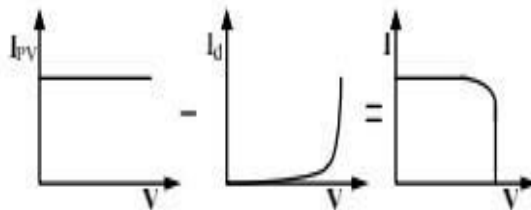


Fig. 9: The PV cell's characteristic I-V curve.

Fig.8 depicts the PV cell equivalent circuit. The following diagram depicts a PV cell as an ongoing source associated in lined up with a diode. The resistance in series is denoted by R_s and the resistance in parallel by R_p . I and V mean the current and voltage, separately, that a PV cell produces. In fig.9, a PV cell's I - V curves. The net cell current I is comprised of the ongoing I_{PV} created by the light and the ongoing I_D generated by the diode. The simulation model and its results for 4-Leg APF with predictive control scheme is shown from fig 10 to 14. The THD variation is shown in fig 15.

VI. SIMULATION RESULTS

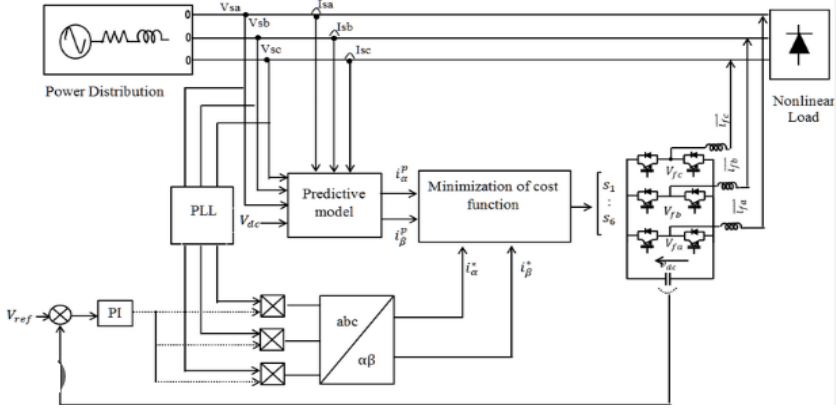
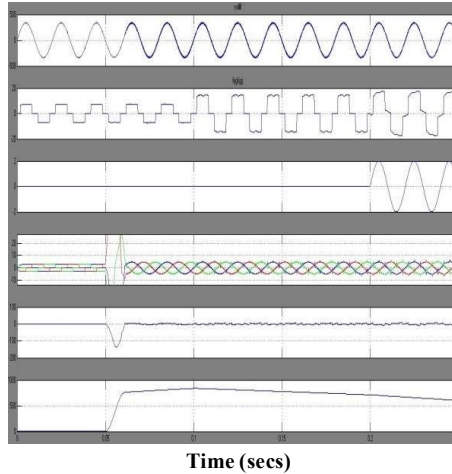
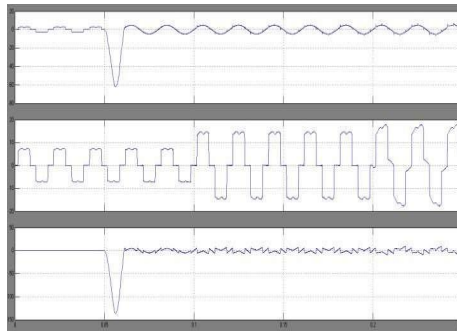


Fig. 10. Predictive control Scheme for 4-Leg APF model



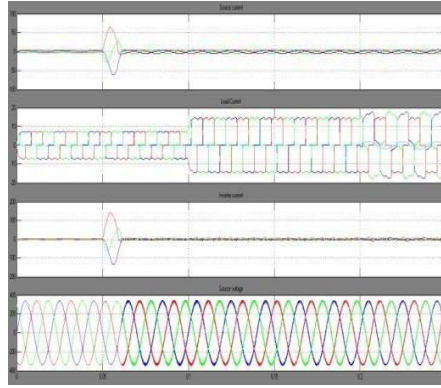
Time (secs)

Fig. 11. Modeled waveforms of the suggested control method. a) Voltage from the source, phase to neutral There are two types of load current: (b) line current and (c) neutral load current. The current has three separate phases, thus (d) is correct. (e) currents that cancel out other undesirable ones. A DC-to-DC voltage converter is the focus of (f).



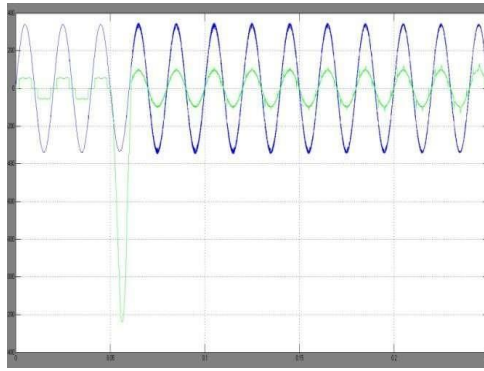
Time (secs)

Fig. 12. Compensating current waveforms for a single-phase source and load, as modelled.



Time (secs)

Fig. 13. Three-phase source load current, neutral line current, compensating current, and source voltage are shown in simulated waveforms.



Time (secs)

Fig. 14. Illustrates the power factor.

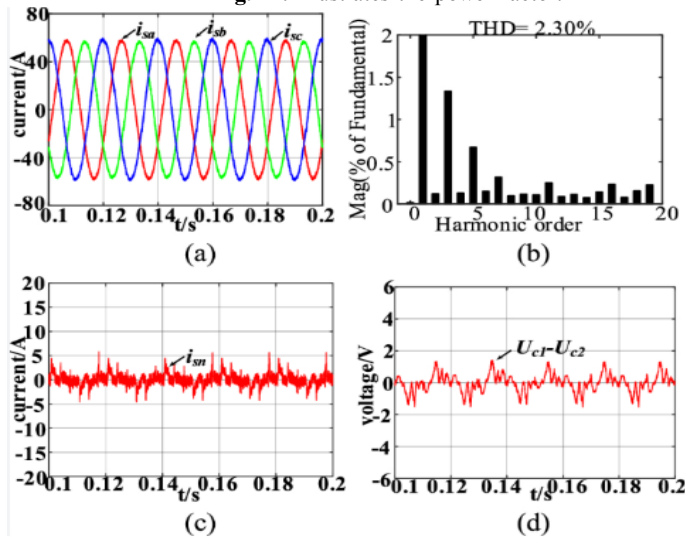


Fig. 15. Total Harmonic Distortion of currents

VII. CONCLUSION

An effective controller is utilized in conjunction with a simulation of a 4-leg inverter-based APF for a photovoltaic (PV) system to enhance power quality. Prescient control of the converter current circle has been exhibited to be a utilitarian reaction for dynamic power channel executions. A three-phase control scheme was presented in this study. For a 3 stage 4 wire framework, the power quality is improved at the purpose in like manner coupling (PCC) using a four-leg grid interfacing inverter. This means that the suggested controller efficiently feeds all fluctuations in dc link actual power onto the main grid. The PV system in this RES uses a active power filter (APF) in view of a four-leg inverter.

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